

Design and Implementation of Axial Flux Induction Motor Single Stator - Single Rotor for Electric Vehicle Application

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Abstract—In the application of a reciprocating compressor failures are often caused by valve malfunction. The failure can be minimized by a proper valve design. This research described an experimental study to investigate characteristics of vibration and cylinder chamber pressure diagrams, as results of open-close valve processes in a single-stage, single-acting air reciprocating compressor. Vibration and pressure were measured at the same crank angle. The modification was materialized by a change in a valve seat profile of a discharge side. The study was conducted for several working loads. Compared to the discharge valve which was (commonly) supplied by the manufacturer, the result showed that a provision of a taper angle on a discharge valve seat would decrease vibration level, made longer discharge opening, and would increase discharge pressure.

Keywords—reciprocating compressor, valve seat, vibration, pressure.

I. INTRODUCTION

Various researches have been done to develop electric vehicles in order to satisfy human needs, as vehicles with fuel in general. One of existing problems in the design of electric vehicles is on the Energy and Power Density. This problem also indirectly related to the size and weight of the vehicle. Along with the vehicle design that more concise and complex, room for electric motor placement in the vehicle is very limited, so electric motor which is efficient in size is needed.

To solve previous problems, there is an idea to make electric motors (especially induction motors) with axial flux method. With Axial Flux Induction Motor (AFIM), the construction of electrical motors that is slimmer than conventional motors become possible, because the stator and rotor facing each other (flux radial method) rather than lengthwise tubular shape of conventional induction motors. By utilizing the axial flux electric motor to drive the car, the installation of the electric motors will be easier.

II. AXIAL FLUX INDUCTION MOTOR

A. AFIM

In general the Axial Flux Induction Motor (AFIM) have the same working principle with induction motor (Radial Flux Induction Motor or often abbreviated RFIM), but the design is very different.

The most fundamental difference is magnetic flux direction. Magnetic flux generated of RFIM system is radial direction for machine axis. Whereas the magnetic flux generated of AFIM system have axial direction for machine axis.

History records the first electric machine using axial flux system with electromagnetic induction principle. In 1821, Faraday introduced primitive disc machine is the forerunner of axial flux machines (AF machines). In 1831, Davenport claimed his hold patent of radial flux

machines (RF engine). The next radial flux motor dominates motors utilization of the world until this time.

B. One stator – One rotor AFIM

Design of one stator and one motor is the simplest design of axial flux induction motor. Form consists of only one piece of the stator and the rotor pieces. Flux starting from the stator moves toward the rotor and then back again to the stator. Because it consists of one stator and one rotor only, the torque produced is also smaller than AFIM using two stator or two rotor systems.

III. MOTOR DESIGN

For design an axial flux induction motor. The first motor parameters must be specified as a target to be achieved. The following motor parameters to be made:

Power = 500 watts

Number of phase (m) = 3

VPH - $p_h = 100/58 \text{ V } Y / \Delta$

Pole (p) = 4

Frequency = 50

Power factor (estimated) = 0.8

Efficiency (estimated) = 0.6

From the target data above we can calculate the requirement specification needs supporters , both mechanically and electrically.

A. Early Initialization motors Dimension

To construct an electric engine we must determine the approximate dimensions/sizes of some parts of the machine as a reference for further calculations. In the picture below which shows the cross section of the stator or rotor without slots of AFIM.

Based on research by Campbell (1974) mentioned that for axial flux induction motor with a comparison between De and Dse that the resulting optimal torque is 0.58. To limit the surface area in the axial flux motors specified outer diameter (Dse) is the first 18cm or 0.18 m . Thus we can get the value of Ds from

$$K_d = \frac{D_s}{D_{se}} = 0,58 \quad (1)$$

It can be calculated :

$$R_{in} \text{ (inner radius)} = 0,5 \times D_s \quad (2)$$

$$R_{out} \text{ (outer radius)} = 0,5 \times D_{se} \quad (3)$$

$$\text{core length (le)} = R_{out} - R_{in} \quad (4)$$

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$$R_{ave} \text{ (average radius)} = \sqrt{(R_{out} \cdot R_{in})} \quad (5)$$

Surface area which produces force / torque (Sr)

$$S_r = \Pi (R_{out}^2 - R_{in}^2) \quad (6)$$

B. Characteristics of Core Materials

In this study using different types of materials on the market and affordable that is steel type ST37. ST37 has the composition of a mixture of iron (Fe) with various other metals according with the following table 1. From the curve in Figure 4 shows that for B at 0 to 1.4 Tesla, material has a permeability (μ) that relatively similar.

C. Selection of Magnetic load

Magnetic load is the maximum flux density (B) that generated by the core to produce a turning force. The flux density will affect the amount of magnetic flux generated by the core. Simply put the amount of flux is the product of the flux density with the flux-producing area.

$$\phi = \int_A B dA \quad (7)$$

Figure 5 below explain that calculation of the flux density at AFIM is operated at each level of diameter, the core is divided into 10 sections circularly n1 n10 with a difference of 0.4 cm radius.

$$B_{ds(1...n)} = \frac{B\delta \cdot \tau_{s(1...n)}}{\tau_{s(1...n)} - bs} \quad (8)$$

$B_{ds(1...n)}$ = flux density at the core
 $B\delta$ = flux density at the air gap
 τ_s = Distance between the stator slot
 bs = width of stator slot

The flux density at the air gap of 0.6 selected in these session, table 2 recapitulate of flux density at the core of stator.

Based on table 2, Flux density at the core position should be located before the saturation. From magnetisation curve, flux density at the core can be obtained by a material before saturation phase, so that flux density at the air gap (B) = 0,6 can be applied.

D. Stator Design

1) Determination The Number of Slot

The Motor use 3 phase source and 4 pole, therefore the minimal number of the pole is $3 \times 4 = 12$ slot. AFIM design use 12 slot and it minimize production motor cost, if it is so many slot then the production cost will so expensive and the complication so high.

2) Stator Winding

Fluks Φ is the product of the flux density (B) with flux-producing area. The wide area here is surface wide area per one pole (1 pole pitch). On draft AFIM has 4 poles so that the area

$$S = S_r / p \quad (9)$$

Kw value Or winding factor is determined by the distribution factor (Kb) and the pitch factor (Kp)

$$K_w = K_b \cdot K_p \quad (10)$$

$$K_b = \frac{\sin m\gamma/2}{m \sin \gamma/2} \quad (11)$$

$$K_p = \cos \theta_{sp}/2 \quad (12)$$

Voltage is induced in the stator phase-neutral voltage is reduced by the loss of copper loss (R), the empirical E_{ind} is 97% of V ph-n. Then be obtained,

$$N = \frac{E_{ind}}{4,44 f \Phi K_w} \quad (13)$$

E. Current Requirement Calculation

In electric load power will increase proportional to the increase in electric current (voltage fixed). Estimates of current required if the star is linked,

$$I = \frac{P}{\sqrt{3} \times V \times \cos \phi \times \eta} \quad (14)$$

F. Requirement torque Analysis and load current

Energy in rotational motion is the result of multiplying the torque by the angular velocity.

$$P = \tau \omega \quad (15)$$

At this axial flux motors angular velocity is

$$n_s = \frac{120 f}{p} \text{ rpm} \quad (16)$$

Taking the estimated slip = 6% then the rotational rotor speed of 1400 rpm.

$$n_r = n_s - (s \times n_s) \quad (17)$$

$$\omega = n_r \times \frac{2\pi}{60} \text{ rad/s} \quad (18)$$

torque is the turning force which directly proportional to the length of the arm, so as torque is multiplication between the resulting force in air gap (tangensial stress), the rotor radius and surface area.

$$\tau = S_r R_{ave} \sigma F_{tan} \quad (19)$$

accordingly,

$$\sigma F_{tan} = \frac{\tau}{R_{ave} S_r} \quad (20)$$

thereafter, with the known specification before, total conductor can be known with the derivative tangensial equation,

$$\sigma F_{tan} = \frac{B \cdot A}{2} \quad (21)$$

Linear current density (A) can be obtained.

Afterward the current which flow in the conductor can be obtained from the A value,

$$Z = 2 \cdot m \cdot N \text{ (total conductor in three phase)} \quad (22)$$

$$D = \text{Average Diameter} = 2 \times R_{ave} \quad (23)$$

$$A = \frac{I \cdot Z}{\pi \cdot D} \quad (24)$$

From the calculation above, the current flows in the stator. The current is equal with the force which is used to generate load torque needed. The current magnetization, current for mechanical torque, mechanical and iron losses are not included in the current value that can be obtained above. The current in this final project is 3,48 A.

G. Rotor Design

The axial flux motors use a cage rotor system, rotor sections has a shape almost the same stator, which is a dish that also has a slot. On the rotor there are two types of material. That is as the aluminum rotor bar and St37 as core. Aluminum with a mixture having resistifitas between $2,82 \times 10^{-8} \Omega m$ lower than the resistivity of steel is $1.18 \times 10^{-7} \Omega m$, whereas the relative permeability of aluminum close to 1. Permeability is very small compared with the relative permeability of the

steel materials worth about 687. Number of slots in the rotor must adjust number of stator slots with the provisions particular to avoid the occurrence the problem as follows :

- Rotor lock
Can be avoided with the provisions of $S_s \neq S_r$.
- Crawling phenomena
Crawling phenomena is when a rotating magnetic field is produced in the air gap and usually nonsinusoidal containing harmonics 5 and 7, Harmonics 7 will result in a dip in the torque and speed characteristics. This phenomena can be avoided with the provisions of $S_s - S_r \neq \pm 3P$, $S_r \neq 6PG$, $S_r < 1.25 S_s$ (S_s = slot stator, rotor S_r = Slot).
- Sound will be heard when spinning rotor (noisy operation)
This can be avoided with the provisions of $S_s - S_r \neq \pm 1, \pm 2, (\pm P \pm 1), (\pm P \pm 2)$.

Of some of the provisions of the above then by the number of stator slots 12 can be found a safe value for the rotor slot 9.

After be known a rotor design, need be done checking a flux density in a rotor core. A Checking is done with an equal method in stator

$$B_{dr(1...n)} = \frac{B\delta \cdot \pi_{(1...n)}}{\pi_{(1...n)} - br} \quad (25)$$

$B_{dr(1...n)}$ = a flux density in rotor core
 $B\delta$ = a flux density in a air gap
 π = distance between a rotor slot
 br = width a rotor slot

The result is as follows:

From a data result above so be ensured that a flux density in rotor core must below a saturation phase ($B > 1.4T$).

H. Yoke

Yoke is a part of motor that function as a flux track back to initial way. Based on yoke position is divided become two type are yoke is put on a stator and yoke is put on a rotor. Yoke position is behind a part stator and rotor appropriate with figure 6. A yoke thickness can be calculated with formula

K_f is a flux density. A flux density at yoke (B) is determined appropriate with a material ability to generating magnet (magnetization curve) without occur a saturation condition. In this matter, be equated with a maximum flux density at stator core is 1,4.

1) Air Gap

Between stator and rotor AFIM is found air gap, distance of this air gap very influence at rotor performance. In empirical, distance of optimal air gap can be calculate with formula as follows

$$L_g = 3,06 \cdot \frac{6560}{D + 2280} \quad (27)$$

D = Stator diameter in millimeters

IV. MOTOR CONSTRUCTION

A. Stator Construction

In the process of making the stator, stator is divided into two parts : stator plate and core. These two parts are made separately. Core is made in three layers, total density of each layer are equal to 20mm. All of this three

layers are attached to stator plate as shown in figure 7. Before assembly process, lamination process is done in the core, this lamination process using various method such as cleaner layers (stainless layer), thermostable paint, and iron glue which has epoxy material. All of this material have isolator characteristic. The purpose of the lamination process is to decrease eddy current loss.

After the process in stator has done, the winding can be place in the slot. The winding is formed before with particular mold, the specifications of the winding are 4 pole double layers, each pole has 50 coils, the dimension of the coil is determined before. Before the winding can be place, it must be coated with the film paper which separate and make intersection of the winding with the core body safe. Thereafter this copper winding is coated with the lamination liquid.

B. Rotor Construction

Rotor construction had similar with stator construction. Both of them have similar forms. The differences, rotor core didn't stand out form a slot but have flat form. Rotor have two parts, rotor bar and rotor core. Rotor bar made of aluminum material, and the rotor core made of steel ST37 materials.

When both of primary component have made, then we can do assembly process. It needs two poles to prop both of components.

V. MOTOR TEST

Equivalent circuit is needed to analyze induction motor. By using this equivalent circuit of induction motor, we can analyze torque curve to speed. To get equivalent circuit of induction motor, we have to do some test, such as: DC test, no load test, and block rotor test.

A. DC Test

DC test is purposed to measure the resistance of stator winding (R_{s1}). At DC test, DC voltage variable as stator winding input for each phase,

$$R_1 = \frac{V}{I} \quad (28)$$

B. DC Test

No load test is purposed to know about the parameters of magnetizing current, core-loss, and to make sure that motor rotate. The Result of Measuring AFIM motor without load is shown in figure 5.

Current has reached nominal value before the nominal voltage reached, for no load test, then the parameter for the no load test taken from one of the test voltage. In this motor, the parameter taken at voltage value 15 V, 5,1 A of current, power factor $\cos\theta = 0,8$.

$$Y = \frac{I_{nl}}{V_{nl}} \quad (29)$$

$$Y = G + jB \quad (30)$$

$$R_c = \frac{1}{G} \quad (31)$$

$$X_m = \frac{1}{B} \quad (32)$$

C. Block Rotor Test

Block rotor tests are performed to fulfill the needs of the equivalent circuit parameters of an induction motor. That X_2 and R_2 values connected in the series equivalent

circuit. In the rotor block tests the results obtained voltage 16 V, current 7A, and power factor = 0.92

$$Z_{br} = \frac{V_{br}}{I_{br} \sqrt{3}} \quad (33)$$

$$Z_{br} = R_{br} + jX_{br} \quad (34)$$

$$R_2' = R_{br} - R_1 \quad (35)$$

$$X_{br} = X_1 + X_2' \quad (36)$$

$$\text{With estimation that } X_1 = X_2' \quad (37)$$

D. Motor Equivalent Circuit

After all parameters has obtained, then equivalent circuit can be arranged, figure 12 shows the single flux axial one stator-one rotor induction motor equivalent circuit.

E. Curve Characteristic of Torque Vs Velocity

Curve of torque vs velocity is using to known caharcteristic of the torque changing to velocity. With this curve the performance and characteristic of power output motor can be know. Curve of torque vs velocity can be determine from differential of torque equation from equivalen circuit motor.

$$V_{thev} = \frac{X_m}{\sqrt{R_1^2 + (X_1 + X_m)^2}} V_1 \quad (38)$$

Because $X_m \gg X_1$ and $X_m + X_1 \gg R_1$ then

$$R_{thev} = R_1 \left(\frac{X_m}{X_1 + X_m} \right)^2 \quad (39)$$

$$X_{thev} \approx X_1 \quad (40)$$

$$\tau_{ind} = \frac{3V_{thev}^2 R_2 / s}{\omega_{sync} [(R_{thev} + R_2 / s)^2 + (X_{thev} + X_2)^2]} \quad (41)$$

After get the torque equation, curve of torque vs velocity can be made. This curve made by insert value of slip(s) begin from 0 to 1, the result of the curve shown below

Depending of characteristic curve torque vs velocity, it gained the maximum torque 0,79Nm at slip 0,07 or 7%. The determined curve from measuring is with nominal voltage per-phase 15 Volt, maximum current 7A. Thus power input and output in this motor can be compared.

$$P_{in} = \sqrt{3} \cdot V_{ph-ph} \cdot I \cdot \cos \theta \quad (42)$$

$$P_{out} = P_{mekanik} = \tau \omega \quad (43)$$

thus the efficiency,

$$\eta = \frac{P_{out}}{P_{in}} \times 100\% \quad (44)$$

VI. CONCLUSION

1. Prototype of one stator-one rotor axial flux induction motor that has been made has total length dimensions 66 mm and outer diameter 200mm. Motor is made from commercially available material that is steel types St 37. By using this axial flux system, induction motor which formed from 2 pieces disc will has smaller dimension than common induction motor (radial flux), so it will very suitable when applied as a driver in electric vehicles.
2. No-load test will conducted at applied voltage of 15 volt, this axial flux induction motor can operated in ideal speed that is 1366 rpm (slip 9%), with flowing current 5,1 A.
3. Equivalent circuit of induction motor obtained from some following testing. With DC voltage test, the stator resistance ($R_1 = 1,14 \text{ ohm}$) is obtained, in no-load test will obtained core losses parameter ($R_c = 2,13 \text{ ohm}$) and the magnetizing reactance ($X_m = 2,85 \text{ ohm}$). In locked-rotor test will obtained leakage reactance value ($X_1 = 0,26 \text{ ohm}$, $X_2 = 0,26 \text{ ohm}$), and rotor resistance ($R_2 = 0,07 \text{ ohm}$).
4. Based on torque vs. speed curve analysis, shown that the maximum torque is 0.79 Nm at speed 146.01 rad /s, so that the maximum output power is 115.34 watts. Starting torque is 0,19 Nm..

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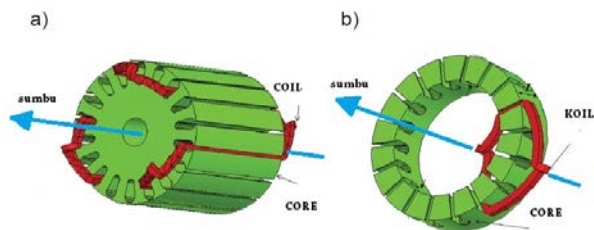


Figure 1. (a) Rotor construction and radial flux directions RFIM. (b) Stator construction and axial flux direction

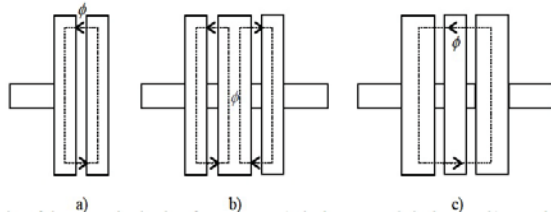


Figure 2. AFIM motor types. a) Single stator - single rotor. b) single stator-double rotor. c) double stator-single rotor

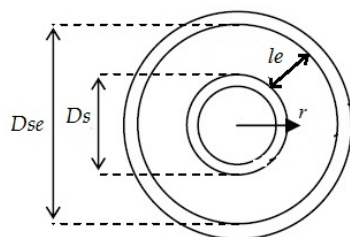


Figure 3. Sectional AFIM

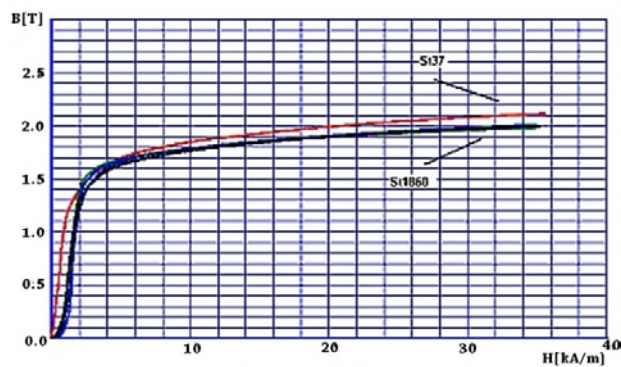


Figure 4. Kurva magnetisasi material St37

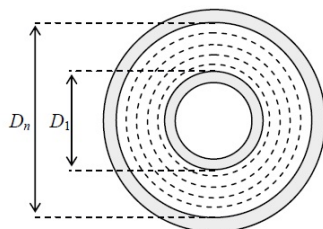


Figure 5. The distribution area of the stator and rotor

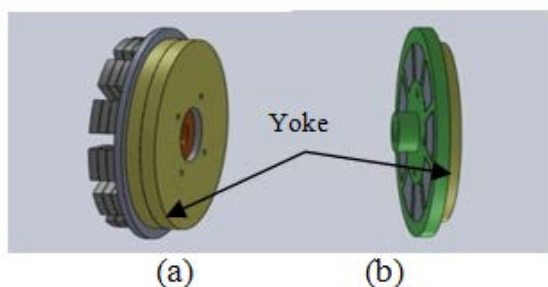


Figure 6. (a) Stator yoke, (b) rotor yoke



Figure 7. Stator Production



Figure 8. Stator with the winding

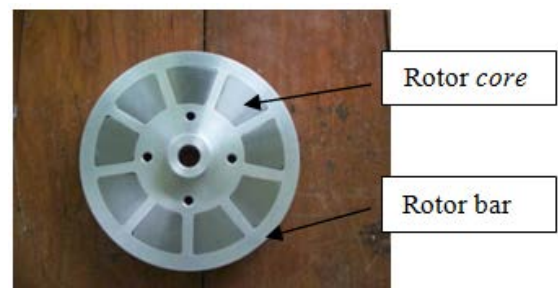


Figure 9. Rotor construction

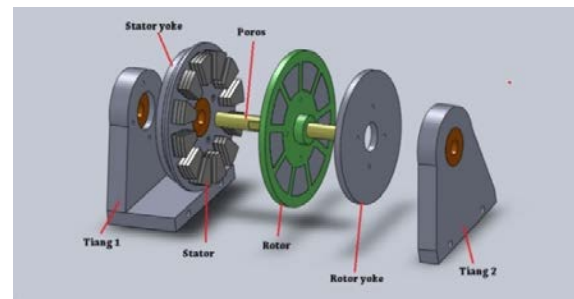


Figure 10. the arrangement of Assembly motor

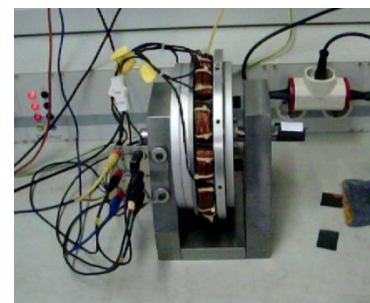


Figure 11. No-Load Test

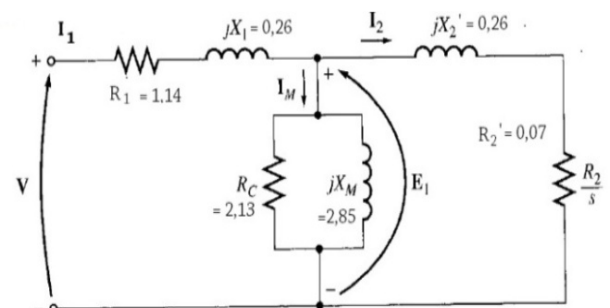


Figure 12. Equivalent circuit of the axial flux motor test results

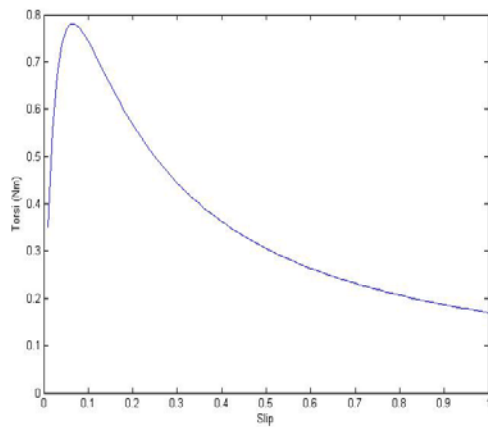


Figure 13. Kurva karakteristik Torsi vs Kecepatan

TABLE 1.
ST37 CONSTITUENT MATERIAL COMPOSITION MIXTURE IN (%)
Material Number 1.0037 - ST37-2/S235JR

C	Si	Mn	P	S	Cr	Mo	Ni	N
Max 0,17	Max 0,30	Max 1,40	Max 0,045	Max 0,045	-	-	-	Maks 0,009

TABEL 2.
CALCULATION RESULTS OF FLUX DENSITY AT STATOR CORE

No	Width of stator slot (m)	Distance between stator slot (m)	B in core (T)
1	0.015	0.027213333	1.336899563
2	0.015	0.029306667	1.229077353
3	0.015	0.0314	1.148780488
4	0.015	0.034016667	1.073269062
5	0.015	0.036633333	1.016024653
6	0.015	0.03925	0.971134021
7	0.015	0.041866667	0.934987593
8	0.015	0.044483333	0.905257207
9	0.015	0.0471	0.880373832

TABLE 3.
THE CALCULATION RESULT OF A FLUX DENSITY IN ROTOR CORE

No	Width a rotor slot (m)	Width between a slot (m)	B in core (T)
1	0.01	0.036284444	0.828271897
2	0.01	0.039075556	0.806358912
3	0.01	0.041866667	0.788284519
4	0.01	0.045355556	0.769704588
5	0.01	0.048844444	0.754462243
6	0.01	0.052333333	0.741732283
7	0.01	0.055822222	0.730940834
8	0.01	0.059311111	0.721676431
9	0.01	0.0628	0.713636364